

University of California, Davis  
Society of Manufacturing Engineers at UC Davis

EEC 110A\*

Electronics Circuits I

Unofficial

## Mock Unofficial Practice Midterm Exam Solutions

\*Disclaimer: This document is a sample final exam of an Electronics Circuits I midterm. It is a mock exam and does not necessarily reflect the format—in the length of the exam, content covered, the protocol, and other aspects—of an actual midterm exam of EEC 110A in University of California, Davis. However, this covers multiple topics that seems to be a halfway agenda of EEC 110A and this document is an attempt to give students extra practice. The problems in this document are written entirely by the author. Any similarity, either in part or in whole, is a complete coincidence. If an error is caught, or if you have any questions and inquiries, please contact the author at [mnhyu@ucdavis.edu](mailto:mnhyu@ucdavis.edu).

A calculator is not encouraged where not needed. Scoring distribution for each question is not provided as it is not graded and discourages students from judging the importance of a topic over another. Please do not physically print out this document.

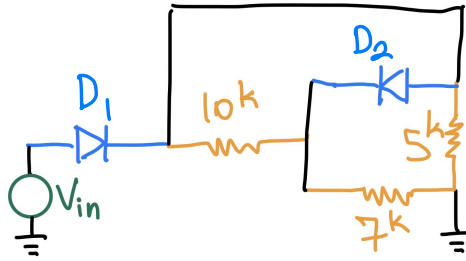
This solution to this examination has eight (8) pages, including this front cover page and the end topology sheet.

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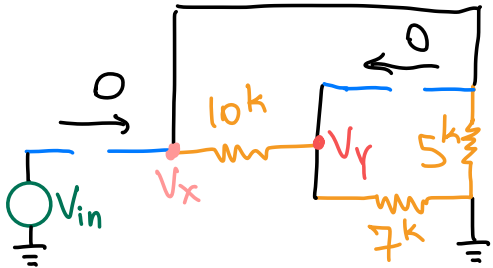
Question 1) Diode Circuits	
Question 2) Small Signal Model for BJTs	
Question 3) Amplifier Topology	

# Solutions 2

- 1) Plot  $P_{IN}$ , the power dissipated by  $V_{IN}$ , with respect to  $V_{IN}$  using the constant voltage model with  $V_{D,ON} = 1V$ . Label your graphs with units and important values.

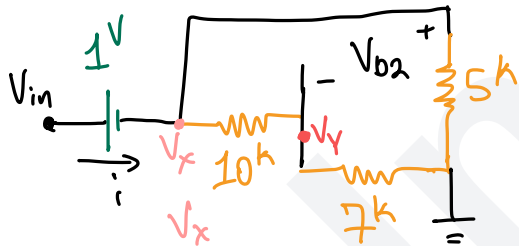


a) Assume  $D_1$  &  $D_2$  off ( $V_{in}$  very negative)



$D_1: V_{D1} = V_{in} - V_x < V_{D,ON}$        $D_2: V_x - V_y < V_{D,ON}$   
 Since no currents are flowing,  
 $V_x = 0$  &  $V_y = 0$   
 $V_{in} < V_{D,ON} = 1 \checkmark$  if  $V_{in}$  very negative  
 $P = 0$  both off

b) Assume  $D_1$  on &  $D_2$  off ( $V_{in} > 1$  and  $V_{in} < TBD$ )



$D_1: i = (V_{in} - V_{D,ON}) / (17k || 5k) > 0$

$1V = V_{in} > V_{D,ON} = 1V \checkmark$

If  $V_{in}$  is just larger than  $1V$

$D_2: V_{D2} < V_{D,ON}$   
 $V_{D2} = (V_{in} - V_{D,ON}) - (V_{in} - V_{D,ON}) \frac{7k}{17k}$

$\frac{10}{17} (V_{in} - V_{D,ON}) < V_{D,ON}$

$\frac{10}{17} V_{in} < \frac{27}{17} V_{D,ON}$

$1 < 2.7 \checkmark$

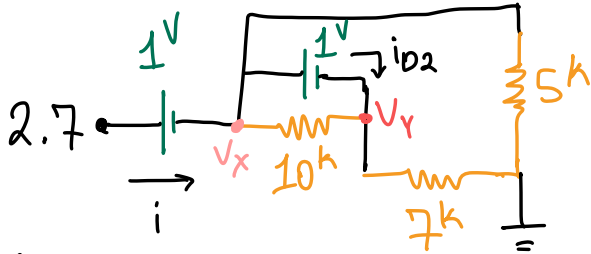
$P_{in} = V_{in} i \Big|_{\substack{D_1 \text{ on} \\ D_2 \text{ off}}} = V_{in} \left( \frac{V_{in} - 1}{17k || 5k} \right)$

Check that  $P_{in}$  @  $V_{in} = 1V$  should be  $0W$

$P_{in} = V_{in} \left( \frac{V_{in} - 1}{17k || 5k} \right) = 1 \left( \frac{1 - 1}{17k || 5k} \right) = 0$

# Solutions 3

c) Assume  $D_1$  &  $D_2$  on ( $V_{in} < 2.7$ )



$$D_1: i = \underbrace{(V_{in} - 2V_{D_{on}})}_{V_y} / 7k + \underbrace{(V_{in} - V_{D_{on}})}_{V_x} / 5k > 0$$

$$= \frac{5V_{in} - 10 + 7V_{in} - 7}{35k} = \frac{12V_{in} - 17}{35k} > 0$$

$$32.4 = 12(2.7) > 17 \checkmark$$

$$D_2: i_{D2} = \underbrace{(V_{in} - 2V_{D_{on}})}_{V_y} / 7k - 1V / 10k > 0$$

$$(V_{in} - 2V_{D_{on}}) 10 > 7$$

$$2.7 - 2 \times 1 > 0.7$$

$$0.7 > 0.7 \checkmark$$

$$P = V_{in} i \Big|_{\text{both on}} = V_{in} \frac{12V_{in} - 17}{35k}$$

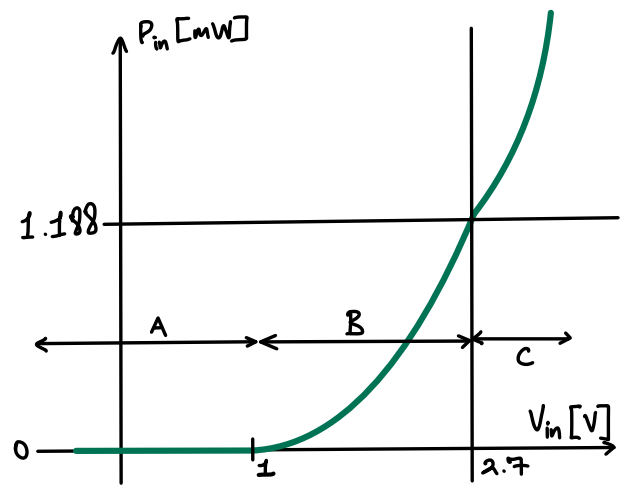
Check  $P_{in}$  @  $V_{in} = 2.7V$  — at boundary of condition B and C should equal

b)  $P_{in} = V_{in} \left( \frac{V_{in} - 1}{17k \parallel 5k} \right) = 2.7 \left( \frac{2.7 - 1}{17k \parallel 5k} \right) = 1.188mW$

c)  $P_{in} = V_{in} \frac{12V_{in} - 17}{35k} = 2.7 \left( \frac{12(2.7) - 17}{35k} \right) = 1.188mW$

Conclusion:

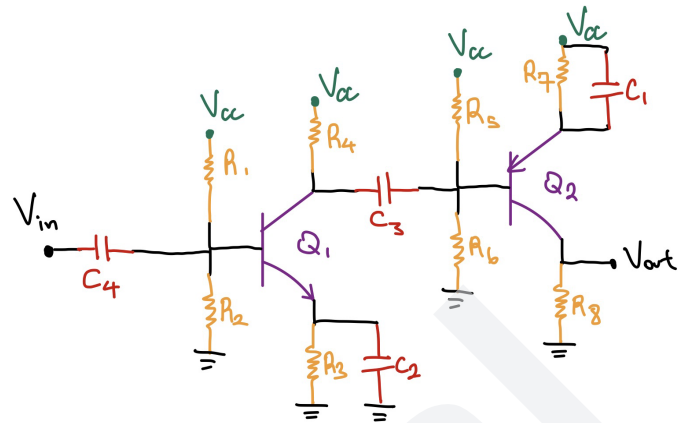
	$D_1$	$D_2$	$P$
a) $V_{in} \leq 1V$	off		0
b) $1V < V_{in} \leq 2.7$	on	off	$V_{in}(V_{in} - 1) / (17k \parallel 5k)$
c) $V_{in} > 2.7$	on		$V_{in}(12V_{in} - 17) / 35k$



Not drawn to scale

# Solutions 4

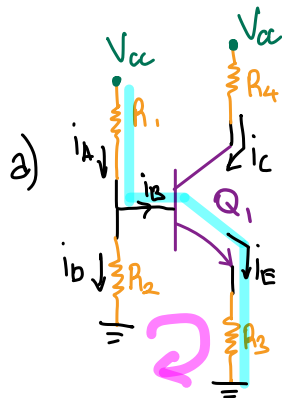
2) You are not allowed to use the topology formula sheet to solve this problem. Given the circuit below with 1 PNP BJT, 1 NPN BJT, 4 capacitors, and 8 resistors, and all transistors operate in the forward active region in the same temperature with  $V_A = \infty$ ,



- Find the transconductance of each BJT in terms of the resistor values, capacitor values,  $\beta_1$  (current gain of  $Q_1$ ),  $\beta_2$  (current gain of  $Q_2$ ),  $V_{BE}$  (for the NPN BJT),  $V_{EB}$  (for the PNP BJT),  $V_T$  (thermal voltage), and  $V_{CC}$ . Do not neglect base currents.
- Using the small signal model, find  $V_{OUT}/V_{IN}$  in terms of  $g_{m1}$ ,  $g_{m2}$ ,  $r_{\pi 1}$ ,  $r_{\pi 2}$ , and other resistor values in the circuit. You do not have to replace  $g_{m1}$  and  $g_{m2}$  with results from part a); instead, simply put  $g_{m1}$  and  $g_{m2}$ . Assume that all capacitors are large enough so that any AC signal would short them.
- Repeat part b), but instead of gain, find input resistance and output resistance (resistance looking into  $V_{IN}$  and  $V_{OUT}$ , respectively).

You could use the topology formula sheet to check your work for part b) and c).

# Solutions 5



( $\beta$ 's below are  $\beta_1$ )

KVLs:

$$V_{cc} - R_1 i_A - V_{BE} - R_3 i_E = 0$$

$$V_{cc} - R_1 (i_B + i_D) - V_{BE} - R_3 i_E = 0$$

$$R_2 i_D = V_{BE} + R_3 i_E$$

$$i_D = \frac{V_{BE} + R_3 i_E}{R_2}$$

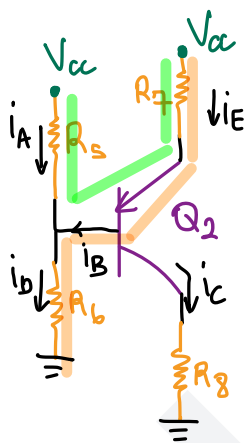
Replace  $i_D$  from magenta into cyan  $\rightarrow V_{cc} - R_1 \left( i_B + \frac{V_{BE} + R_3 i_E}{R_2} \right) - V_{BE} - R_3 i_E = 0$

Recall  $i_E = (\beta + 1) i_B \rightarrow V_{cc} - R_1 \left( i_B + \frac{V_{BE}}{R_2} + \frac{R_3}{R_2} (\beta + 1) i_B \right) - V_{BE} - R_3 (\beta + 1) i_B = 0$

Isolate  $i_B \rightarrow V_{cc} - R_1 \frac{V_{BE}}{R_2} - i_B \left( R_1 \left( 1 + \frac{R_3}{R_2} (\beta + 1) \right) + R_3 (\beta + 1) \right) - V_{BE} = 0$

$$i_B = \frac{V_{cc} - V_{BE} \left( \frac{R_1}{R_2} + 1 \right)}{R_1 \left( 1 + \frac{R_3}{R_2} (\beta + 1) \right) + R_3 (\beta + 1)}$$

$$g_{m1} = \frac{\beta_1}{V_T} \frac{V_{cc} - V_{BE} \left( \frac{R_1}{R_2} + 1 \right)}{R_1 \left( 1 + \frac{R_3}{R_2} (\beta + 1) \right) + R_3 (\beta + 1)}$$



( $\beta$ 's below are  $\beta_2$ )

KVLs:

$$V_{cc} - i_A R_5 = V_{cc} - i_E R_7 - V_{EB}$$

$$i_A R_5 = i_E R_7 + V_{EB}$$

$$(-i_B + i_D) R_5 = (\beta + 1) i_B R_7 + V_{EB}$$

$$V_{cc} - i_E R_7 - V_{EB} - i_D R_6 = 0$$

$$i_D = \frac{V_{cc} - i_E R_7 - V_{EB}}{R_6}$$

Replace  $i_D$  from tan into green  $\rightarrow -i_B R_5 + \frac{R_5}{R_6} (V_{cc} - R_7 (\beta + 1) i_B - V_{EB}) = (\beta + 1) i_B R_7 + V_{EB}$

Recall  $i_E = (\beta + 1) i_B \rightarrow -i_B R_5 - \frac{R_5}{R_6} R_7 (\beta + 1) i_B - (\beta + 1) i_B R_7 = V_{EB} - \frac{R_5}{R_6} (V_{cc} - V_{EB})$

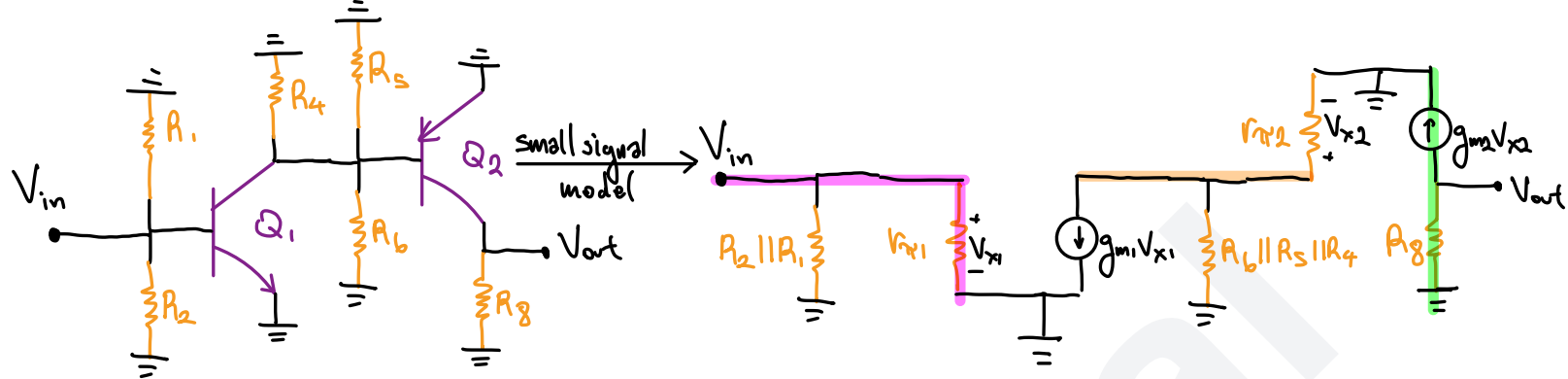
(Isolate  $i_B$ )

$$i_B = \frac{V_{EB} - \frac{R_5}{R_6} (V_{cc} - V_{EB})}{-R_5 - \frac{R_5}{R_6} R_7 (\beta + 1) - (\beta + 1) R_7} = \frac{\frac{R_5}{R_6} V_{cc} - V_{EB} \left( 1 + \frac{R_5}{R_6} \right)}{R_5 + R_7 (\beta + 1) \left( 1 + \frac{R_5}{R_6} \right)}$$

$$g_{m2} = \frac{\beta_2}{V_T} \frac{\frac{R_5}{R_6} V_{cc} - V_{EB} \left( 1 + \frac{R_5}{R_6} \right)}{R_5 + R_7 (\beta + 1) \left( 1 + \frac{R_5}{R_6} \right)}$$

# Solutions 6

b) Shorting capacitors and grounding DC sources



**KVL:**  $V_{in} = V_{x1}$

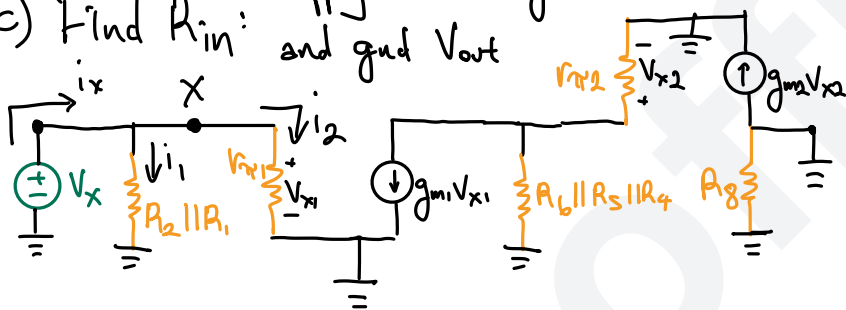
**KCL:**  $g_{m2} V_{x2} = -V_{out}/R_8$

**KCL:**  $g_{m1} V_{x1} = -V_{x2}/(R_6 || R_5 || R_4 || r_{o2})$

$$g_{m1} V_{in} = V_{out} / [(R_6 || R_5 || R_4 || r_{o2}) A_6 g_{m2}]$$

$$\frac{V_{out}}{V_{in}} = (R_6 || R_5 || R_4 || r_{o2}) A_6 g_{m2} g_{m1}$$

c) Find  $R_{in}$ : Apply test voltage at  $V_{in}$  and gnd  $V_{out}$



KCL at X:

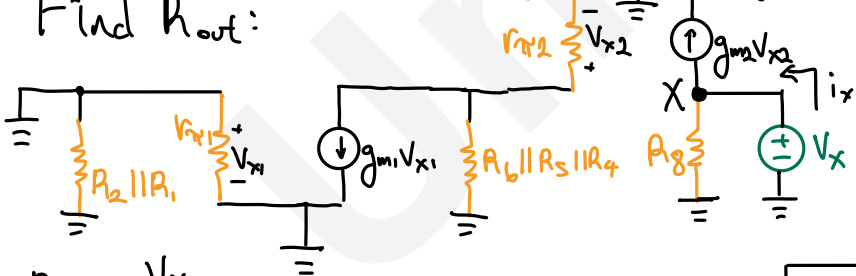
$$i_x = i_1 + i_2 = \frac{V_x}{R_1 || R_2} + \frac{V_x}{r_{\pi 1}}$$

$$R_{in} = \frac{V_x}{i_x} = R_1 || R_2 || r_{\pi 1} = R_{in}$$

If  $\beta_1$  is very large:  $R_{in} \approx R_1 || R_2$

Find  $R_{out}$ :

Apply test voltage at  $V_{out}$  and gnd  $V_{in}$

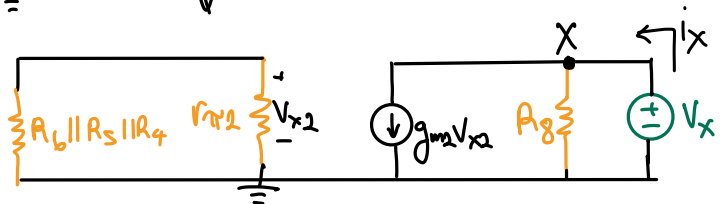


redrawing, ignoring  $R_2 || R_1$  &  $r_{\pi 1}$  b/c they are shorted, & therefore  $V_{x1} = 0$ , so can also ignore  $g_{m1} V_{x1}$

$$R_{out} = \frac{V_x}{i_x}$$

KCL at X:  $i_x = g_{m2} V_{x2} + V_x/R_8$

$$R_{out} = \frac{V_x}{V_x/R_8} = R_8 = R_{out}$$



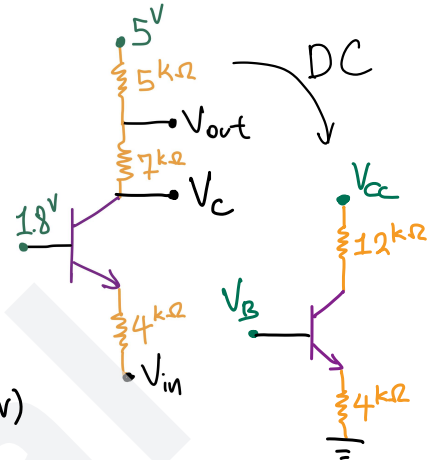
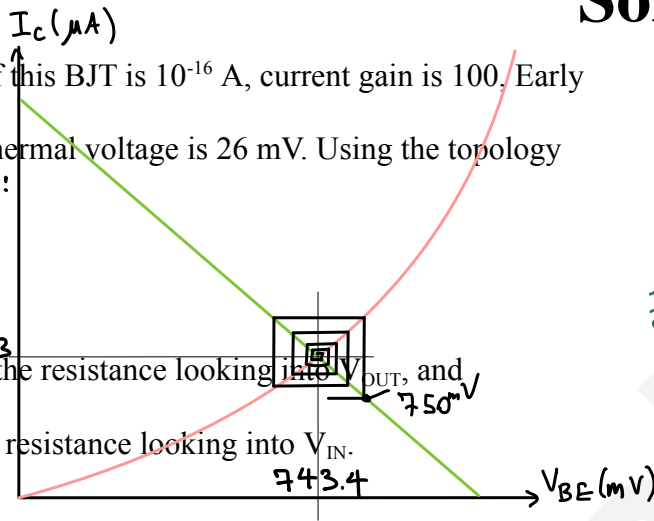
# Solutions 7

- 3) The saturation current of this BJT is  $10^{-16}$  A, current gain is 100. Early voltage is infinite, and thermal voltage is 26 mV. Using the topology

Recursive visual:  
formula sheet, find

- a)  $A_V = V_{OUT}/V_{IN}$ ,  
b) output resistance,  $R_{OUT}$ , the resistance looking into  $V_{OUT}$ , and  
c) input resistance,  $R_{IN}$ , the resistance looking into  $V_{IN}$ .

Do not neglect base currents.



$$I_C = I_S \exp\left(\frac{V_{BE}}{V_T}\right) \rightarrow V_{BE} = V_T \ln \frac{I_C}{I_S} = 0.026 \ln \frac{I_C}{10^{-16}} \quad \therefore I_C = 261.53 \mu A$$

$$KVL: V_B - V_{BE} = I_C \left(\frac{\beta+1}{\beta}\right) 4k$$

$$\frac{100(1.8 - V_{BE})}{101} = I_C = \frac{1.8 - V_{BE}}{4040}$$

Recursive solving: start w/  $V_{BE} = 750$  mV (typical value)

$$I_C^* = 259.9 \mu A$$

$$V_{BE}^* = 0.026 \ln \frac{259.9 \mu A}{10^{-16} \mu A} \approx 743.24 \text{ mV}$$

$$I_C^* = 261.57 \mu A \rightarrow V_{BE}^* = 743.41 \text{ mV}$$

$$I_C^* = 261.53 \mu A \rightarrow V_{BE}^* = 743.40 \text{ mV}$$

close enough! Now... are we in FAR?

$$\frac{V_C}{V_{in}} = \frac{12k\Omega}{\frac{1}{0.01} + 4k} \approx 2.927$$

$$V_{out} = V_C \frac{R_1}{R_2 + R_1} = V_C \frac{5}{12}$$

FAR?

$$V_{BE} < V_{CE} \rightarrow V_B < V_C$$

$$V_B < V_{CC} - (R_1 + R_2) I_C$$

$$I_C (R_1 + R_2) < V_{CC} - V_B$$

$$261.53 \mu A (12k\Omega) < 5 - 1.8$$

$$3.138 < 3.2 \checkmark$$

$$g_m = \frac{I_C}{V_T} = \frac{0.26153 \text{ mA}}{26 \text{ mV}} \approx 0.01 \text{ S} \text{ or } \approx 10 \text{ mS}$$

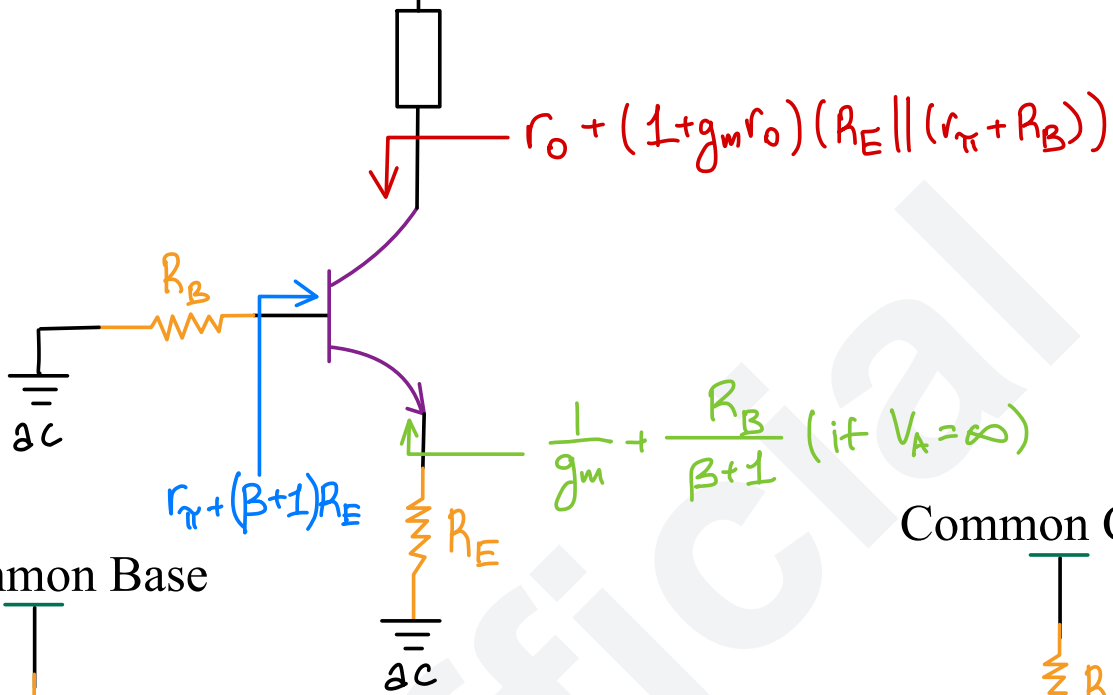
$$\frac{V_{out}}{V_{in}} = \frac{5}{12} 2.927 \approx 1.220 \approx A_V$$

$$R_{in} = \frac{1}{g_m} + 4k \approx 4099.4 \Omega \approx R_{in}$$

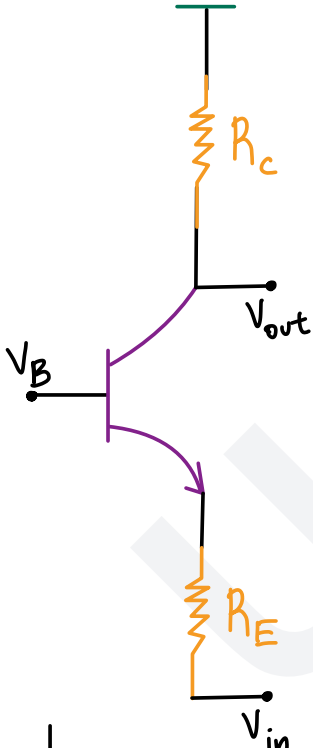
$$R_{out} = 5k \parallel (7k + \infty) = 5k \approx R_{out}$$

# Solutions 8

Topology Sheet

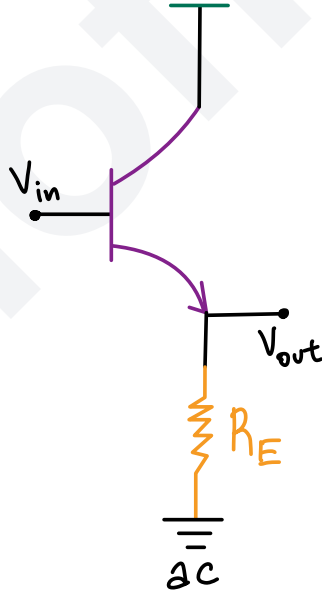


Common Base



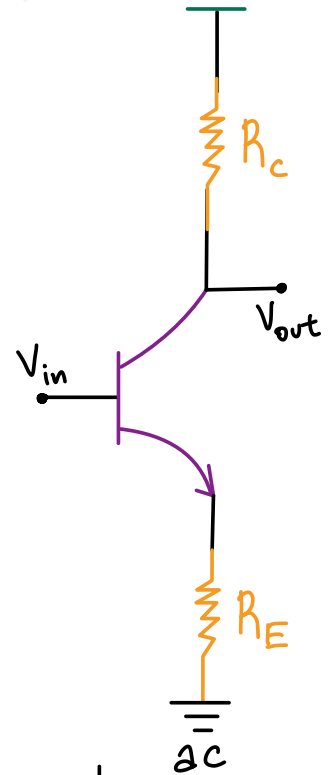
$$\left. \frac{V_{out}}{V_{in}} \right|_{V_A = \infty} = \frac{R_C}{\frac{1}{g_m} + R_E}$$

Emitter Follower



$$\frac{V_{out}}{V_{in}} = \frac{R_E \parallel r_o}{\frac{1}{g_m} + R_E \parallel r_o}$$

Common Collector



$$\left. \frac{V_{out}}{V_{in}} \right|_{V_A = \infty} = \frac{-R_C}{\frac{1}{g_m} + R_E}$$